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Urazhdin, Sergei; Tabor, Phillip; Tiberkevich, Vasil; and Slavin, Andrei, "Fractional Synchronization of Spin-Torque Nano-Oscillators" (2010). Faculty Scholarship. 172.
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Fractional Synchronization of Spin-Torque Nano-Oscillators

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(Received 14 June 2010; published 31 August 2010)

We experimentally demonstrate a series of fractional synchronization regimes (Devil’s staircase) in a
spin-torque nano-oscillator driven by a microwave field. These regimes are characterized by rational
relations between the driving frequency and the frequency of the oscillation. An analysis based on the
phase model of auto-oscillator indicates that fractional synchronization becomes possible when the
driving signal breaks the symmetry of the oscillation, while the synchronization ranges are determined
by the geometry of the oscillation orbit. Measurements of fractional synchronization can be utilized to
obtain information about the oscillation characteristics in nanoscale systems not accessible to direct
imaging techniques.

DOI: 10.1103/PhysRevLett.105.104101 PACS numbers: 05.45.Xt, 75.75.Jn, 76.50.+g, 85.70.Ec

Downscaling of electronic devices requires ever increasing operating current densities. The out-of-equilibrium
state produced by large currents can lead to new phenomena, providing unique opportunities for novel
device architectures. In particular, high current densities in nanoscale magnetic devices result in large-amplitude magneti-
zonization oscillations, forming the basis for a novel type of nanoscale microwave oscillator, the spin-torque nanoo-
sillator (STNO) [1–4]. STNO is not only one of the smallest known auto-oscillators, but also a complex non-
linear dynamical system whose frequency is dependent on the amplitude of the oscillation, due to the dependence of the
demagnetizing fields on the oscillation trajectory.

The nonlinearity of STNO allows one to tune the generation frequency by electric current, and also enhances its ability to synchronize to periodic external signals [5], providing a potential route for enhancing the oscillation characteristics by mutual synchronization in arrays of
STNO. Following this approach, synchronization of two
[6,7] and even four [8] STNO with similar frequencies has been recently observed experimentally.

All the previous experiments on the STNO synchronization utilized a driving signal provided by a microwave current whose frequency \( f_e \) was close to the STNO oscillation frequency \( f_0 \) [9–11]. In this Letter, we utilize a driving signal provided by a microwave field to experimentally demonstrate that the complex nonlinear oscillation characteristics of STNO lead to a large class of synchroni-
zonation phenomena observed not only for \( f_e \) close to \( f_0 \), but also when their ratio \( r = f_e / f_0 \) is close to integer or certain rational numbers. We also show theoretically that these synchronization phenomena provide important information about the properties of the oscillation.

We used e-beam lithography to fabricate devices with
structure Cu(40)Py(15)Cu(8)Py(3.5)Cu(60), where thicknesses are in nm and Py = Ni_{80}Fe_{20}. The free Py(3.5) layer
was patterned into a 100 nm \( \times \) 50 nm nanopillar, while the polarizing Py(15) layer was left extended with dimensions of several micrometers. Magnetic oscillations of the nano-
pillar were induced by a dc bias current \( I_0 > 0 \) flowing from the polarizer to the free layer. The microwave driving
field \( h_e \) was generated by a 300 nm \( \times \) 250 nm Cu micro-
strip antenna fabricated on top of the nanopillar. It was oriented at 45° with respect to the nanopillar easy axis, and
electrically isolated from STNO by a SiO_{2}(50) layer. The Oersted field produced by the microstrip was calibrated by comparing the dependence of \( f_0 \) on the bias field \( H_0 \) to its dependence on the dc current applied to the microstrip. By applying a microwave current to the microstrip, microwave fields of up to \( h_e = 30 \) Oe were generated at the location of
STNO without noticeable heating. Our technique enabled spectroscopic measurements at arbitrary relations between
the driving and the oscillation frequencies, and provided strong driving essential for the observation of the phe-
nomena described below. The parasitic coupling between the microstrip and the STNO was less than \(-20 \) dB, result-
ing in induced microwave currents that did not exceed
40 \( \mu \)A. The bias field was \( H_0 = 350 \) Oe unless specified otherwise, applied in-plane perpendicular to the micro-
strip. All measurements were performed at 5 K. The re-
ported behaviors were confirmed for three devices.

The characterization of STNO was performed by mea-
surements of auto-oscillation spectra. The peaks exhibited a typical full width at half maximum (FWHM) of 5 MHz and Lorentzian line shape characteristic of thermal broad-
ingen [Fig. 1(a)] [12,13]. The oscillation frequency increased with increasing \( H_0 \), and decreased with increasing \( I_0 \) [Fig. 1(b)], consistent with the established properties of similar STNOs [3].

In spectroscopic measurements of oscillation at finite driving field \( h_e \), the oscillation frequency \( f_0 \) remained close to the free-running frequency \( f_0 = 2.94 \) GHz for
most of the values of the driving frequency $f_e$, with the exception of the regions where the ratio $r = f_e/f_0$ is close to all integer ($r = 1, 2, 3, 4$) and several rational ($r = 3/2, 7/3, 5/2, 7/2$) values (Fig. 2). In these regions, $f'_0$ follows a linear relationship with $f_e$, as expected for the synchronized oscillation. Since the spectroscopic peaks observed at $r = 1$ are dominated by the parasitic coupling to the driving signal, in this regime the synchronization can be identified by the breaks in the $f'_0 = \text{const}$ line. In contrast, spectroscopic measurements of all the other synchronization regimes were performed below the driving frequency, eliminating the effects of parasitic signals on the spectra, and enabling us to directly determine the characteristics of the synchronized oscillations.

An example of spectroscopic measurements for one of these regimes ($r = 2$) is shown in Fig. 3(a). The synchronized oscillation peaks become significantly narrower and their maximum spectral amplitude dramatically increases, while the oscillation frequency follows an exact relationship $f'_0/f_e = 2$ within the measurement precision of better than 0.05% [Fig. 3(b)]. To enable a wide frequency scan shown in Fig. 3(a), these measurements were performed with a bandpass of 300 kHz that exceeded the width of the synchronized peak. Separate measurements at our setup’s ultimate resolution of 1 kHz showed that synchronization resulted in an increase of the maximum spectral amplitude by three orders of magnitude $f_e$, and a similar magnitude of decrease of FWHM to the smallest value of 1.4 kHz [Fig. 3(d)]. These results demonstrate that the oscillator becomes fully phase locked with the driving signal for at least 1 ms ($3 \times 10^6$ oscillation cycles) without a single phase slip. We note that the total power generated by STNO is determined by the amplitude of the magnetization oscillation, which is not significantly affected by the driving field. Indeed, the increase of the peak amplitude is compensated by the decrease of the linewidth, so that the total generated power remains approximately constant.

To quantitatively characterize the synchronization, we define the synchronization interval $\Delta f_s = (f_{e,\max} - f_{e,\min})/r$, where $f_{e,\max}$ ($f_{e,\min}$) is the maximum (minimum) driving frequency at which the synchronization is observed. Although both the integer interval $\Delta f_2$ and the fractional interval $\Delta f_{5/2}$ increase with the driving amplitude, the value of $\Delta f_2$ depends linearly on $h_e$, while the dependence of $\Delta f_{5/2}$ on $h_e$ is strongly nonlinear [Fig. 4(a)]. Integer synchronization persisted at the smallest driving signal $h_e = 1.3$ Oe used in our measurements, while the $r = 5/2$ synchronization disappeared at $h_e < 4$ Oe, demonstrating that strong driving provided by our technique is essential for the observation of fractional synchronization. The value of $\Delta f_2$ increases at small $I_0$, exceeding 1 GHz at $I_0 < 0.9$ mA. In contrast, $\Delta f_{5/2}$ decreases at small $I_0$, vanishing at $I_0 < 0.9$ mA [Fig. 4(b)].

The complexity of the observed synchronization patterns suggests that they contain intricate information about
The linear dependence of \( k \) obtain an equation for linear dependence of higher orders in \( n \). Let's consider the classic model describing the state of the oscillator only by the phase \( \phi \) defined to increase linearly with time in the autonomous regime \([14,15]\).

Under the influence of a periodic driving signal, the dynamics can be described by

\[
d\phi/dt = \omega_0 + \mu \text{Re}[g(\phi)e^{-i\omega_0 t}]
\]

Here, \( \mu \) is the amplitude of the driving force, \( \omega_{0e} = 2\pi f_{0e} \), and the complex function \( g(\phi) \) determined by the geometry of the oscillation trajectory and the driving signal describes the position-dependent sensitivity of the oscillator to the driving force. Since \( g(\phi) \) is periodic, it can be expressed as \( g(\phi) = \sum_n g_n e^{i n \phi} \), where the Fourier components \( g_n \) contain the information about the oscillation trajectory.

To determine the relation of the function \( g(\phi) \) to synchronization, we consider a fractional synchronized state \( r = p/q \), where \( p \) and \( q \) are mutually simple integers. The oscillator completes \( q \) cycles during \( p \) periods of the driving force, i.e., its phase satisfies \( \phi(t + p/f_0) = \phi(t) + 2\pi q \). Expanding it into a Fourier series, \( \phi(t) = (\phi(0) + \omega_t t + p + \sum_{n \neq 0} \phi_n e^{-i \omega_0 t/p} \) and inserting into Eq. (1), we obtain an equation for \( \phi_n \) which can be solved iteratively in powers of \( \mu \). The condition of solvability of this equation determines the synchronization interval \( \Delta \omega_0 \).

To the first order in \( \mu \), only the synchronization at integer \( r = p/1 \) is possible, with the interval

\[
\Delta \omega_p = 2\mu |g_p|.
\]

Fractional synchronization \( (q > 1) \) appears only in the higher orders in \( \mu \), with intervals \( \Delta \omega_r \propto (\mu g_p)^q \). In particular, for \( r = p/2 \)

\[
\Delta \omega_{p/2} = \frac{\mu^2}{\omega_0} \left| \sum_n g_n g_{p-n} \right|.
\]

The linear dependence of \( \Delta \omega_p \) on \( \mu \approx h \), and the nonlinear dependence of \( \Delta \omega_{p/2} \) are in agreement with Fig. 4(a).

It is generally difficult to calculate the function \( g(\phi) \) that determines the synchronization intervals, since it depends both on the form of the driving signal and the oscillation trajectory. However, it is possible to experimentally determine each component \( g_p \) by measuring the corresponding synchronization intervals [Eq. (2)]. Moreover, symmetry analysis described below enables one to determine the general conditions for the existence of specific synchronization regimes, providing a qualitative insight into the mechanism of fractional synchronization.

The oscillation trajectory of STNO is approximately symmetric with respect to a half-period rotation \( \phi \rightarrow \phi + \pi \) \([3]\). A microwave field parallel to the oscillation symmetry axis \( s \) produces a symmetric driving force; i.e., a \( \pi \) phase shift of the driving signal does not change the oscillation phase. In this case, all the odd Fourier components of \( g(\phi) \) vanish, \( g_{2n+1} = 0 \), and only even integer synchronization regimes can be observed. On the other hand, \( h \perp s \) produces an antisymmetric driving force. In this case, all the even components of \( g(\phi) \) vanish, \( g_{2n} = 0 \), and only odd integer synchronization regimes can be observed.

To understand the implications of our symmetry analysis for fractional synchronization, we consider the half-integer regime [Eq. (3)]. Since the numbers \( p \) and \( q = 2 \) are mutually simple, \( p \) is odd. Therefore, the indices of the Fourier components \( g_n \) and \( g_{p-n} \) in Eq. (3) have different parity. Since the even spectral components describe perturbations that are symmetric with respect to a half-period rotation, and the odd ones describe antisymmetric perturbations, the \( q = 2 \) fractional synchronization of a symmetric auto oscillator is possible only if the driving signal contains both symmetric and antisymmetric components, or in other words, if the signal breaks the symmetry of the oscillator. In our experiment, \( h \) was parallel to \( H_0 \), at 45° with respect to the nanopillar axis. Modeling shows that, as a result of the demagnetizing effects, \( s \) was rotated by 20° with respect to \( h \), creating the symmetry-breaking geometry necessary for the fractional synchronization.

To confirm the results of our analysis, we performed numerical simulations of STNO synchronization to a microwave field (Fig. 5). The temporal evolution of magnetization was determined by integration of the Landau-Lifshitz equation within a macrospin approximation, which included the spin-torque and the driving field terms, with the parameter values corresponding to the experimental conditions of Fig. 2: \( H_0 = 350 \) Oe oriented at 45° with the nanopillar easy axis, \( I_0 = 1.3 \) mA, \( h = 13 \) Oe, saturation magnetization \( M_s = 640 \) G, easy-axis anisotropy field \( H_a = 200 \) Oe, Gilbert damping constant \( \alpha = 0.01 \), and dimensionless spin-polarization efficiency \( \eta = 0.18 \). The spectral characteristics of the driven oscillation were determined by Fourier transformation of the simulated dynamics.

In agreement with our symmetry analysis, only the even integer synchronization regimes are prominent for \( h \parallel s \) [Fig. 5(a)], and only the odd integer synchronization regimes are prominent for \( h \perp s \) [Fig. 5(a)]. In both cases, small but finite fractional synchronization ranges are
caused by the slight asymmetry of the oscillation trajectory. On the other hand, simulation of the experimental configuration $\mathbf{h}_s \parallel \mathbf{H}_0$ [Fig. 5(c)], corresponding to an angle of $20^\circ$ between $\mathbf{h}_s$ and $\mathbf{s}$, produces large integer and fractional synchronization ranges, in excellent semi-quantitative agreement with the data (Fig. 2). Additional simulations for $f_e < f_0$ showed that fractional synchronization to symmetry-breaking perturbations is also possible at $r < 1$.

Our symmetry analysis is equally applicable to synchronization by microwave current. In this case, the symmetry of the driving force is determined by the direction of the current polarization. In the published measurements of the main synchronization regime $r = 1$ [9–11], the current polarization was nearly parallel to $\mathbf{s}$, and consequently the driving force was approximately symmetric. Since synchronization with $r = 1$ requires antisymmetric perturbation [see Eq. (2)], our analysis explains why a large magnitude of the driving current comparable to $I_0$ was required in these measurements. In contrast, a significantly more efficient $r = 2$ synchronization can be expected in this geometry.

In conclusion, we have experimentally demonstrated synchronization of a spin torque nano-oscillator at fractional ratios between the frequency of the driving microwave field and the frequency of the oscillation. The phenomenon of fractional synchronization opens new routes for efficient synchronization of auto-oscillators whose frequencies are not close to each other, and for the development of novel nanoscale signal processing devices such as microwave frequency converters. We developed a general model of synchronization, which shows that fractional synchronization of symmetric oscillators becomes possible only if the driving force breaks the oscillation symmetry, and more generally, that the efficiency of synchronization in any regime is determined by the symmetry of the driving signal. Fractional synchronization for a controlled symmetry of the driving signal represents a novel tool for the characterization of auto-oscillators. For instance, one can determine the orientation of the precession axis by measuring the dependence of the synchronization intervals on the direction of the driving field. This characterization technique is especially important for the nanoscale oscillators such as STNO, which are not accessible to the standard imaging and characterization techniques.

This work was supported by NSF Grants DMR-0747609, ECCS-0653901, and ECCS-0967195, the Research Corporation, U.S. Army TARDEC, RDECOM W56HZV-09-P-L564, and W56HZV-10-P-L687.